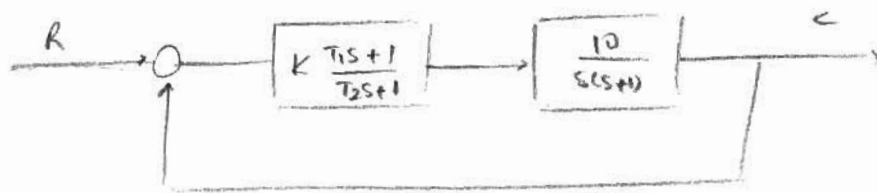


MAE 4310 SOLUTION TO HW # 7

B.7.7)



$$\frac{C(s)}{R(s)} = \frac{K \cdot \frac{T_1 s + 1}{T_2 s + 1} \cdot \frac{10}{s(s+1)}}{1 + K \frac{T_1 s + 1}{T_2 s + 1} \cdot \frac{10}{s(s+1)}}$$

$$= \frac{10K(T_1 s + 1)}{(s^2 + s)(T_2 s + 1) + 10K T_1 s + 10K}$$

Char. poly. = $T_2 s^3 + (1 + T_2) s^2 + (10K T_1 + 1) s + 10K$

Desired dominant closed loop poles: $\zeta = 0.5, \omega_n = 3 \text{ rad/sec}$

$$s^2 + 2\zeta\omega_n s + \omega_n^2 = s^2 + 3s + 9$$

$$T_2 s^3 + (1 + T_2) s^2 + (10K T_1 + 1) s + 10K = (s^2 + 3s + 9)(as + b)$$

To get a stable closed loop

$$s = -\frac{b}{a} < 0 = \frac{b}{a} > 0$$

(2)

$$(s^2 + 3s + 9)(as + b) = as^3 + (3a + b)s^2 + (3b + 9a)s + 9b.$$

$$\left. \begin{array}{l} T_2 = a \\ T_2 + 1 = 3a + b \end{array} \right\} \frac{3a + b}{a} = \frac{T_2 + 1}{T_2} \quad \frac{b}{a} + 3 = 1 + \frac{1}{T_2}$$

$$10kT_1 + 1 = 3b + 9a$$

$$10k = 9b.$$

$$\boxed{\frac{b}{a} = \frac{1}{T_2} - 2}$$

from the condition $\frac{b}{a} > 0$

$$\frac{1}{T_2} - 2 > 0$$

$$\boxed{T_2 < 1/2}$$

for $T_2 = 1/3$

$a = 1/3, b = 1/3 \Rightarrow$ other pole is located at $\boxed{s = -1}$

$$\boxed{k = 0, 3.}$$

$$, \boxed{T_1 = 1.}$$

(3)

B-7-8)

closed-loop transfer function

$$\frac{C(s)}{R(s)} = \frac{k(Ts+1)}{s(s+2) + k(Ts+1)}$$

Since the closed-loop poles are located at $s = -2 \pm j2$

$$s(s+2) + k(Ts+1) = (s+2+j2)(s+2-j2)$$

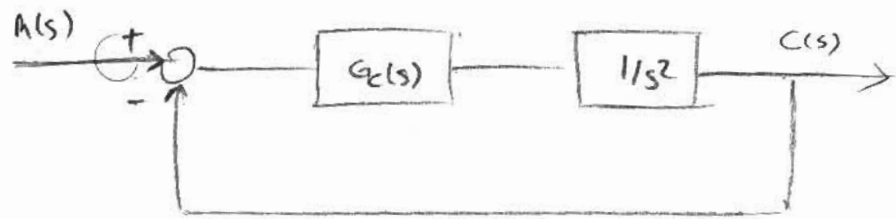
$$s^2 + (2+kT)s + k = s^2 + 4s + 8$$

$$2+kT = 4 \quad ; \quad k = 8$$

$$T = 0.25$$

B.7-10)

(4)



Dominant desired-closed loop poles $s = -1 \pm 1j$

$$(s+1+j)(s+1-j) = s^2 + 2s + 2$$

$$G_c(s) = k \frac{s+a}{s+b}$$

$$\frac{C(s)}{R(s)} = \frac{k \frac{s+a}{s+b} \cdot 1/s^2}{1 + k \frac{s+a}{s+b} \cdot \frac{1}{s^2}}$$

$$= \frac{k(s+a)}{s^3 + bs^2 + ks + ka}$$

$$\text{char. poly} = s^3 + bs^2 + ks + ka$$

$$s^3 + bs^2 + ks + ka = (s^2 + 2s + 2)(\alpha s + \beta)$$

Similar to B.7-7 $\frac{\beta}{\alpha} > 0$

$$s^3 + bs^2 + ks + ka = \alpha s^3 + (2\alpha + \beta)s^2 + (2\alpha + 2\beta)s + 2\beta$$

$$\left. \begin{array}{l} 1 = \alpha \\ b = 2\alpha + \beta \end{array} \right\} \frac{2\alpha + \beta}{\alpha} = b \Rightarrow \frac{\beta}{\alpha} = b - 2$$

$$k = 2(\alpha + \beta)$$

$$ka = 2\beta$$

from the condition $\frac{\beta}{\alpha} > 0$

$$b - 2 > 0 \quad b > 2$$

pole @ $s = -\frac{\beta}{\alpha}$ should be located far enough from the dominant poles.

for $b = 5$

$$\alpha = 1, \quad \beta = 3$$

$$s = -3$$

$$k = 8$$

$$\alpha = 6/8 \Rightarrow$$

$$a = 0.75$$

Thus,

$$G_c(s) = 8 \frac{s + 0.75}{s + 5}$$

MATLAB CODE

```

num=[0 0 1];
den=[1 0 1];
numc=[8 6];
denc=[1 5 8 6];
t=0:0.02:10;
c1=step(num,den,t);
c2=step(numc,denc,t);
plot(t,c1,'-',t,c2,'-')
grid
title('Unit step responses of uncompensated and compensated systems')
xlabel('t(sec)')
ylabel('Outputs')
text(1.9,0.85,'Compensated system')
text(4.1,1.65,'Uncompensated system')

```

